

Conditional Variable-Length Coding for Gray-Level Pictures

H. GHARAVI*

(Manuscript received May 25, 1983)

This paper presents a new variable-length coding algorithm for bit-rate reduction of digital images. The coding scheme is based on the fact that by taking into account the local conditional statistics of the picture, the coding efficiency can be improved considerably. Results of computer simulations are presented in terms of picture quality and required bit rate. Simulations show that for various Differential Pulse Code Modulation (DPCM) coded pictures, coding efficiency is such that a bit rate below the entropy of the DPCM quantizer output is attained. For a 3-bit DPCM encoded picture, a bit rate of 1.64 bits/picture element is achieved.

I. INTRODUCTION

Two popular methods for reducing bit rates of pictures are Differential Pulse Code Modulation (DPCM)¹ and transform coding.² When these methods are applied, their reconstructed signals are not exactly the same as the original Pulse Code Modulation (PCM) input because of the quantization noise. Consequently, these encoders are often referred to as nonreversible encoders. Bit-rate reduction can also be achieved by reversible encoders. These methods are based on the fact that it is not necessary to assign an equal number of bits to quantized samples that are not equiprobable. Therefore, it may be advantageous to design variable-length code words in which the length of the code

*AT&T Bell Laboratories; present affiliation Bell Communications Research, Inc.

Copyright © 1984 AT&T. Photo reproduction for noncommercial use is permitted without payment of royalty provided that each reproduction is done without alteration and that the Journal reference and copyright notice are included on the first page. The title and abstract, but no other portions, of this paper may be copied or distributed royalty free by computer-based and other information-service systems without further permission. Permission to reproduce or republish any other portion of this paper must be obtained from the Editor.

words are matched to the Probability Distribution Function (PDF) of the source by using very simple procedures. One such procedure has been devised by Shannon³ and Fano and another by Huffman.⁴ Another reversible coding method is run-length coding, in which a sequence of identical symbols is replaced by a code indicating the length and symbol of the sequence. Run-length codes are useful whenever the long runs are highly probable. This method has been widely used for compression of bilevel pictures.⁵

With the above-mentioned reversible coding methods, the entropy obtained from the PDF of a digital image gives a target for the minimum bit rate necessary to describe the source, ignoring the conditional statistics of the picture. If, by some efficient scheme, conditional statistics can be considered, it may be possible to lower this target to a figure that can be described by conditional entropy. One such method has recently been devised by Gharavi et al. for encoding the gray-level pictures using CCITT one-dimensional run-length codes.⁶ In this scheme the run-length statistics of the picture are changed in such a way as to provide a signal that has long bit runs.

In this paper we introduce a similar strategy by developing a scheme in which the picture can be directly coded by the variable-length codes. The code words are designed according to the local conditional statistics of the picture using the Huffman procedure. For example, the picture element (pel) values that are expected to occur most frequently, given the local PDF, are assigned to the shortest code words in accordance. The mathematical model and the details of the coding algorithm are given in the following section.

II. BASIC MODEL

The n -bit PCM coded picture can be described mathematically as a random sequence of 2^n possible outcomes with certain statistical properties, which are given in terms of probabilities. From these probabilities we can calculate the entropy of one pel, which represents a lowest theoretical bound for the one-dimensional memoryless model. The entropy of pictures can be presented as:

$$H_{\text{pel}} = - \sum_{i=1}^{2^n} P_i \text{Log}_2 P_i,$$

where P_i is the probability of the i th outcome (i th level). In the case of an m th-order Markov model, the luminance level of each pel, x_i , is considered to be dependent on the level of m previous surrounding pels. Therefore, the conditional entropy based on the m previous elements is given by:

$$H_c = - \sum_{x_{i-m}}^{2^n} \cdots \sum_{x_{i-1}}^{2^n} \sum_{x_i}^{2^n} P(x_{i-m} x_{i-m-1} \cdots x_i) \cdot \text{Log}_2 P(x_i/x_{i-1}, x_{i-2} \cdots x_{i-m}), \quad (1)$$

where $P(x_{i-m} x_{i-m-1} \cdots x_i)$ is the joint probability of $x_i \cdots x_{i-m}$ and $P(x_i/x_{i-1}, x_{i-2} \cdots x_{i-m})$ is the conditional probability of x_i , given m previous pels $x_{i-1}, x_{i-2} \cdots x_{i-m}$. Equation (1) can be rewritten as

$$H_c = - \sum_{x_{i-m}}^{2^n} \cdots \sum_{x_{i-1}}^{2^n} P(x_{i-m}, x_{i-m-1} \cdots x_{i-1}) \cdot \sum_{x_i}^{2^n} P(x_i/x_{i-1}, x_{i-2} \cdots x_{i-m}) \text{Log}_2 P(x_i/x_{i-1}, x_{i-2} \cdots x_{i-m}) \quad (2)$$

and, on defining H_L as a local entropy,

$$H_L = - \sum_{x_i}^{2^n} P(x_i/x_{i-1}, x_{i-2} \cdots x_{i-m}) \text{Log}_2 P(x_i/x_{i-1}, \cdots x_{i-m}). \quad (3)$$

We express the conditional entropy as

$$H_c = \sum \sum \cdots \sum P(x_{i-m}, x_{i-m-1} \cdots x_{i-1}) H_L. \quad (4)$$

Equation (4) shows that the conditional entropy depends not only on the local entropy but also on its distribution over $(2^n)^m$ independent states. This model leads to the new coding scheme proposed here.

2.1 Coding algorithm

The first step is to use the m previous pels to construct a table of $(2^n)^m$ independent states of picture statistics. For each state, a set of code words is designed according to the probability distribution of the 2^n outcomes of the quantized values. These code words can be derived by employing optimum coding procedures, such as Huffman, or Shannon and Fano, in which a shorter code is assigned to a more probable level. Then, for a given state, each incoming pel is coded by selecting an appropriate code word corresponding to its quantized value.

III. DPCM CODING

In the previous section it is apparent that the total number of states is heavily dependent on the number of quantization levels. For example, for input pictures of 256 levels corresponding to 8 bits/pel, there are 256^m (m is the number of reference elements) states. This figure is rather high, and a simple reduction of the number of quantizing levels would lead to annoying contouring effects. Thus, to convert an 8-bit picture into fewer bits while maintaining approximately the same

quantizer noise, a DPCM encoder is employed. In the DPCM encoder used here, a linear predictor constructs its predicted pel as the weighted summation of previous pels (see Fig. 1); thus

$$\bar{x} = 0.5A + 0.25(B + C), \tag{5}$$

where \bar{x} is the prediction of the present pel. The prediction error is then quantized by a symmetric seven-level quantizer with the transfer characteristics given in Fig. 2. Applying the above coder to the 512×512 , 8-bit/pel picture shown in Fig. 3 results in the seven-level DPCM coded picture shown in Fig. 4.

3.1 Subsampling and interpolation

Subsampling at a factor of 2:1 gives approximately a 2:1 bit-rate reduction. This results in some degradation of pictures, which can be minimized by an efficient interpolation scheme. Figure 5 shows the 2:1 horizontal subsampling pattern employed here. The interpolation is based on averaging the luminance levels of the four surrounding pels. The subsampled picture is DPCM encoded, in which the present pel D (see Fig. 5) is predicted by

$$\text{Prediction of } D = 0.3B + (C + F)0.35.$$

Figure 6 shows the interpolated picture where 2:1 subsampling was employed.

IV. COMPUTER SIMULATIONS

Experiments were carried out by means of computer simulation. Figure 7 shows the transmitter and receiver block diagrams. In our

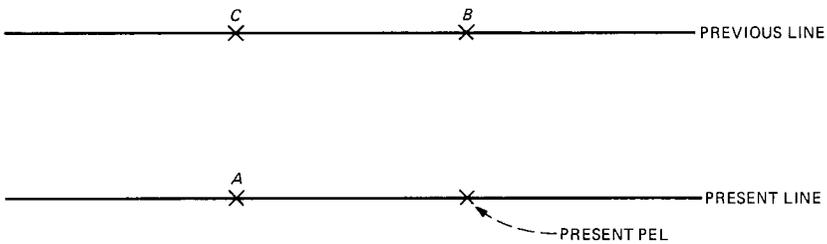


Fig. 1—Configuration of pels used for prediction.

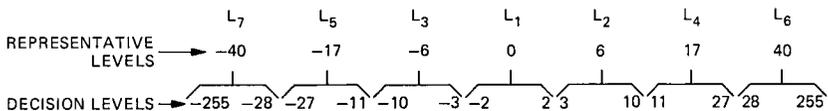


Fig. 2—Transfer characteristics of the quantizer.



Fig. 3—Original picture.

simulation, for a set of seven-level DPCM coded pictures, tables of 49 and 343 states were constructed by observing the quantized values of two and three previous elements, respectively (A, B and A, B, C of Fig. 8). As a matter of convenience only, the table of 49 states with the corresponding code words is shown in Table I. For each state, the seven code words are obtained from the probability distribution function of the seven outcomes, using the standard Huffman procedure.

As Fig. 7a shows, the input signal is first DPCM encoded to generate a 3-bit DPCM coded picture. The coded picture is then applied to the conditional variable coder, which uses the line and sample delay to represent the values of neighbors B and A , respectively. These values, along with current pel value, comprise the input for the encoder Look Up Table (LUT). The encoder LUT uses this information to give the bits of the appropriate variable-length code word, as shown in Table I. The output is smoothed by a buffer to allow for the uniform bit rate in the transmission channel. In the receiver, the transmitted variable-length code words are received from the channel and applied to a buffer and then to the decoder LUT (see Fig. 7b). The conditional variable-length decoder LUT also receives the decoded sample and line-delayed signals to represent the values of neighbors B and A , respectively. Based on this information, the decoder determines



Fig. 4—Three-bit DPCM coded picture.

x	o	x^C	o	x^F
o	x^B	c^A	x^D	o
x	o	x^E	o	x
o	x	o	x	o

X : SAMPLE SELECTED FOR TRANSMISSION
 O : SAMPLE DROPPED

INTERPOLATION OF $A = (B+C+D+E) / 4$

Fig. 5—Subsampling pattern used for interpolation and prediction.

whether the current pel code word is the one expected to occur most frequently, second most frequently, and so on, as specified in Table I. The output of these is then applied to the DPCM decoder, which serves as the receiver output signal.



Fig. 6—Subsampled and interpolated picture.

4.1 Results

Table II presents the bit-rate results of the variable-length encoder with the corresponding entropies and conditional entropies. The results of variable-length coding based on the statistics of the quantizer output (ignoring the conditional statistics) are also included in Table II as a reference.

By looking at the tabulated results, it is clear that a considerable improvement is obtained by using this conditional variable-length algorithm compared with a nonconditional variable-length encoder. An even better improvement is obtained by increasing the number of reference elements from two to three (third row of Table II) at the expense of complexity, as the number of states is increased from 49 to 343. In the case of 2:1 subsampling, a 2:1 bit-rate reduction is unobtainable. This is due to the lower spatial correlation as a result of an increased distance between neighboring pels, as compared to a non-subsampled case. However, the most important conclusion from Table II is that bit rates below the entropy of the DPCM quantizer output are achievable.

V. CONCLUSION

A new concept for a variable-length coding scheme has been pre-

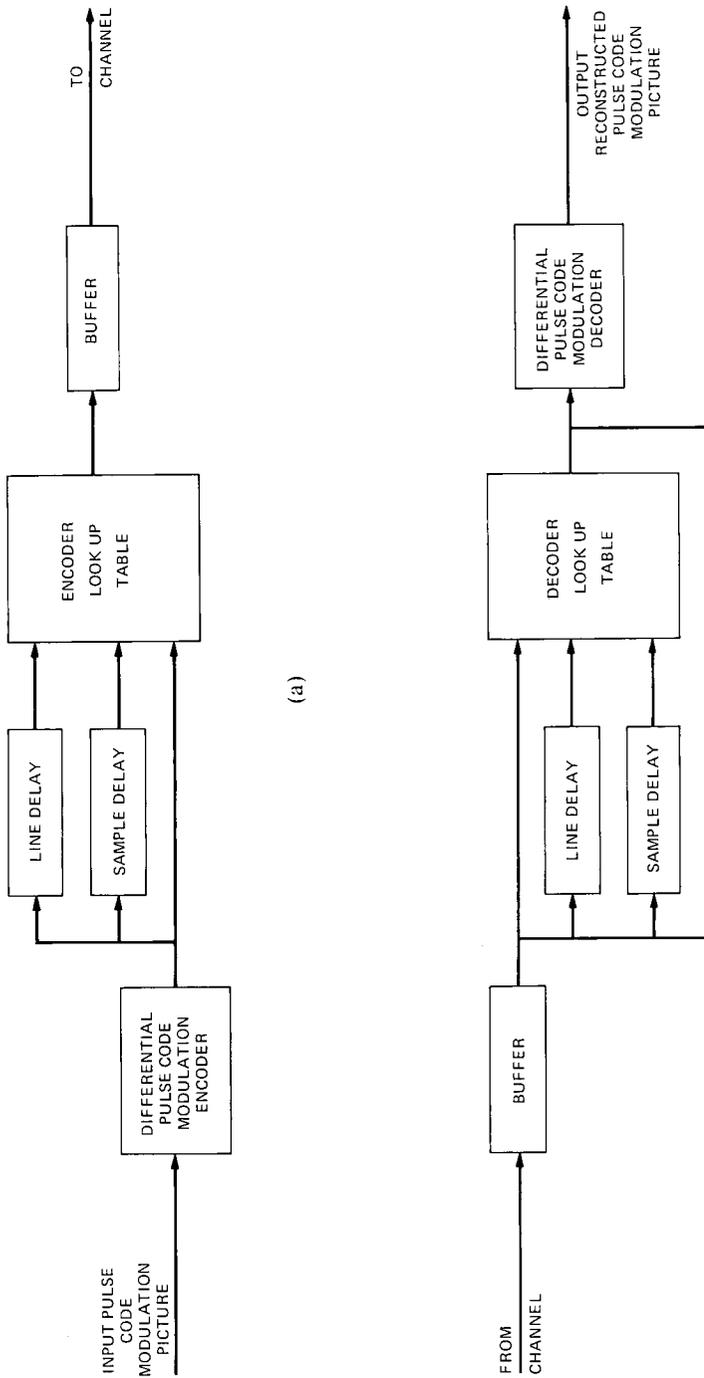


Fig. 7—Block diagram of (a) transmitter and (b) receiver.

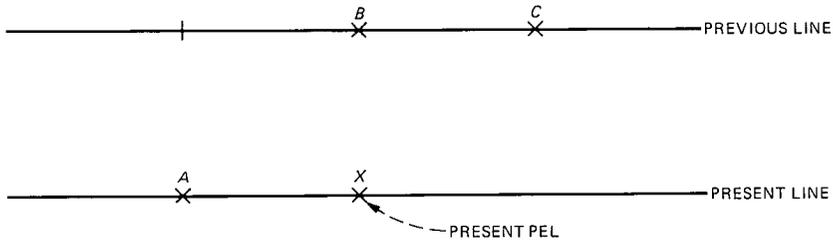


Fig. 8—Configuration of reference elements used for conditional variable-length coding.

Table I—Table of variable-length code words

State	Neighbors		Code Words						
	A	B	L ₁	L ₂	L ₃	L ₄	L ₅	L ₆	L ₇
1	L ₁	L ₁	L ₁ 1	L ₂ 01	L ₃ 001	L ₄ 00001	L ₅ 000001	L ₆ 000001	L ₇ 000000
2	L ₁	L ₂	L ₁ 0	L ₂ 11	L ₃ 101	L ₄ 1001	L ₅ 10001	L ₆ 100001	L ₇ 100000
3	L ₁	L ₃	L ₁ 0	L ₃ 11	L ₂ 101	L ₅ 1001	L ₄ 10001	L ₆ 100001	L ₇ 100000
4	L ₁	L ₄	L ₂ 11	L ₁ 10	L ₄ 01	L ₃ 001	L ₇ 0001	L ₆ 00001	L ₈ 00000
5	L ₁	L ₅	L ₃ 11	L ₅ 10	L ₁ 01	L ₂ 001	L ₆ 0001	L ₄ 00001	L ₇ 00000
6	L ₁	L ₆	L ₅ 1	L ₆ 01	L ₃ 0011	L ₂ 0010	L ₇ 0000	L ₁ 00011	L ₄ 00010
7	L ₁	L ₇	L ₇ 1	L ₁ 01	L ₂ 0011	L ₆ 0010	L ₅ 0000	L ₃ 00011	L ₄ 00010
8	L ₂	L ₁	L ₁ 0	L ₃ 11	L ₂ 101	L ₄ 1001	L ₅ 10001	L ₇ 100001	L ₆ 100000
9	L ₂	L ₂	L ₁ 0	L ₂ 11	L ₃ 101	L ₄ 1001	L ₅ 10001	L ₇ 100001	L ₆ 100000
10	L ₂	L ₃	L ₃ 0	L ₁ 11	L ₂ 101	L ₅ 1001	L ₄ 10001	L ₆ 100001	L ₇ 100000
11	L ₂	L ₄	L ₂ 11	L ₄ 10	L ₁ 01	L ₃ 001	L ₇ 0001	L ₅ 00001	L ₆ 00000
12	L ₂	L ₅	L ₃ 11	L ₅ 10	L ₁ 01	L ₂ 001	L ₆ 0001	L ₄ 00001	L ₇ 00000
13	L ₂	L ₆	L ₆ 0	L ₅ 01	L ₇ 0011	L ₃ 0010	L ₄ 0000	L ₁ 0011	L ₂ 00010
14	L ₂	L ₇	L ₇ 0	L ₄ 01	L ₆ 0011	L ₁ 0010	L ₂ 0000	L ₃ 0011	L ₅ 00010
15	L ₃	L ₁	L ₁ 0	L ₂ 11	L ₃ 101	L ₄ 1001	L ₅ 10001	L ₆ 100001	L ₇ 100000
16	L ₃	L ₂	L ₁ 0	L ₂ 11	L ₃ 101	L ₄ 1001	L ₅ 10001	L ₇ 100001	L ₆ 100000
17	L ₃	L ₃	L ₁ 0	L ₃ 11	L ₂ 101	L ₅ 1001	L ₄ 10001	L ₆ 100001	L ₇ 100000
18	L ₃	L ₄	L ₂ 0	L ₁ 11	L ₄ 101	L ₃ 1001	L ₇ 10001	L ₅ 100001	L ₆ 100000
19	L ₃	L ₅	L ₃ 11	L ₁ 10	L ₅ 01	L ₂ 001	L ₆ 0001	L ₄ 00001	L ₇ 00000
20	L ₃	L ₆	L ₆ 0	L ₅ 01	L ₃ 0011	L ₂ 0010	L ₁ 0000	L ₇ 00011	L ₄ 00010
21	L ₃	L ₇	L ₇ 0	L ₆ 111	L ₄ 110	L ₃ 100	L ₂ 1010	L ₅ 10111	L ₁ 10110

Table I—Continued

State	Neighbors		Code Words						
	A	B							
22	L ₄	L ₁	L ₃ 0	L ₁ 11	L ₂ 101	L ₅ 1001	L ₄ 10001	L ₇ 100001	L ₆ 100000
23	L ₄	L ₂	L ₂ 11	L ₁ 10	L ₃ 01	L ₄ 001	L ₅ 0001	L ₇ 00001	L ₆ 00000
24	L ₄	L ₃	L ₃ 0	L ₁ 11	L ₅ 101	L ₂ 1001	L ₄ 10001	L ₇ 100001	L ₆ 100000
25	L ₄	L ₄	L ₄ 0	L ₂ 11	L ₁ 101	L ₇ 1001	L ₃ 10001	L ₅ 100001	L ₆ 100000
26	L ₄	L ₅	L ₅ 0	L ₃ 111	L ₁ 110	L ₂ 100	L ₆ 1010	L ₄ 10111	L ₇ 10110
27	L ₄	L ₆	L ₆ 11	L ₇ 10	L ₅ 01	L ₄ 001	L ₂ 0001	L ₁ 00001	L ₃ 00000
28	L ₄	L ₇	L ₇ 0	L ₄ 11	L ₂ 101	L ₃ 1001	L ₅ 10001	L ₁ 100001	L ₅ 100000
29	L ₅	L ₁	L ₂ 11	L ₁ 10	L ₃ 01	L ₅ 001	L ₄ 0001	L ₆ 00001	L ₇ 00000
30	L ₅	L ₂	L ₂ 0	L ₁ 111	L ₄ 110	L ₃ 100	L ₅ 1010	L ₆ 10111	L ₇ 10110
31	L ₅	L ₃	L ₃ 11	L ₁ 10	L ₂ 01	L ₅ 001	L ₄ 0001	L ₆ 00001	L ₇ 00000
32	L ₅	L ₄	L ₄ 11	L ₂ 10	L ₇ 01	L ₁ 001	L ₃ 0001	L ₅ 00001	L ₆ 00000
33	L ₅	L ₅	L ₅ 11	L ₃ 10	L ₆ 01	L ₁ 001	L ₂ 0001	L ₄ 00001	L ₇ 00000
34	L ₅	L ₆	L ₆ 0	L ₅ 11	L ₃ 101	L ₁ 1001	L ₂ 10001	L ₇ 100001	L ₄ 100000
35	L ₅	L ₇	L ₆ 0	L ₇ 11	L ₄ 101	L ₅ 1001	L ₂ 10001	L ₁ 100001	L ₃ 100000
36	L ₆	L ₁	L ₁ 11	L ₆ 10	L ₇ 011	L ₃ 010	L ₂ 000	L ₅ 0011	L ₄ 0010
37	L ₆	L ₂	L ₆ 11	L ₇ 01	L ₂ 00	L ₄ 1011	L ₅ 1010	L ₁ 1001	L ₃ 1000
38	L ₆	L ₃	L ₁ 0	L ₆ 110	L ₂ 101	L ₃ 1111	L ₅ 1110	L ₇ 1001	L ₄ 1000
39	L ₆	L ₄	L ₇ 11	L ₆ 10	L ₄ 01	L ₂ 001	L ₁ 0001	L ₃ 00001	L ₅ 00000
40	L ₆	L ₅	L ₅ 11	L ₆ 10	L ₃ 01	L ₁ 001	L ₂ 0001	L ₄ 00001	L ₇ 00000
41	L ₆	L ₆	L ₆ 1	L ₅ 01	L ₃ 0011	L ₇ 0010	L ₁ 0000	L ₄ 00011	L ₂ 00010
42	L ₆	L ₇	L ₆ 0	L ₇ 11	L ₅ 101	L ₄ 1001	L ₂ 10001	L ₃ 100001	L ₁ 100000
43	L ₇	L ₁	L ₇ 0	L ₅ 110	L ₃ 101	L ₄ 1111	L ₆ 1110	L ₂ 1001	L ₁ 1000
44	L ₇	L ₂	L ₇ 11	L ₂ 10	L ₃ 011	L ₄ 010	L ₁ 000	L ₅ 0011	L ₆ 0010
45	L ₇	L ₃	L ₇ 0	L ₅ 111	L ₆ 110	L ₄ 100	L ₃ 1010	L ₁ 10111	L ₂ 10110
46	L ₇	L ₄	L ₄ 11	L ₇ 10	L ₂ 01	L ₃ 001	L ₁ 0001	L ₆ 00001	L ₅ 00000
47	L ₇	L ₅	L ₇ 11	L ₆ 10	L ₅ 011	L ₃ 010	L ₄ 000	L ₂ 0011	L ₁ 0010
48	L ₇	L ₆	L ₇ 1	L ₆ 01	L ₄ 000	L ₅ 00111	L ₃ 00110	L ₂ 00101	L ₁ 00100
49	L ₇	L ₇	L ₇ 1	L ₄ 01	L ₂ 000	L ₁ 00111	L ₃ 00110	L ₆ 00101	L ₅ 00100

Table II—Performance of variable-length coding algorithms for 512 x 512 pictures

Variable-Length Encoder	DPCM Encoder	Entropy of Quantizer Output (bits/pel)	Conditional Entropy (bits/pel)	Bit Rate (bits/pel)
Based on quantizer output	Nonsampled	1.84	—	1.94
Based on conditional statistics ref. elements A, B	Nonsampled	1.84	1.64	1.73
Based on conditional statistics ref. elements A, B, C	Nonsampled	1.84	1.53	1.64
Based on conditional statistics ref. elements A, B	Sampled	1.05	0.92	1.01
Based on conditional statistics ref. elements A, B, C	Sampled	1.05	0.89	0.97

sented in which the structure of the variable-length code word changes from pel to pel, depending on the local conditional statistics of the picture. In this scheme 3-bit DPCM words are coded by assigning a variable-length code word to represent each pel as a function of its expected frequency of occurrence, given the neighbor pel values (state). We have shown that for sampled and nonsampled 3-bit DPCM coded pictures, bit rates of 0.97 bits/pel and 1.64 bits/pel, respectively, are achievable. We therefore conclude that by using the conditional variable-length encoder it is possible to obtain bit rates below the entropy of the quantizer output.

REFERENCES

1. H. Gharavi, "Bandwidth Compression of Digital Color Television Signals Using Block-Adaptive DPCM," *IEE Proc. E. Commun., Radar & Signal Proc.*, 127, No. 5 (October 1980), pp. 405-9.
2. A. N. Netravali and J. O. Limb, "Picture Coding—A Review," *Proc. IEEE*, 68, No. 3 (March 1980), pp. 366-406.
3. C. E. Shannon, "A Mathematical Theory of Communication," *B.S.T.J.*, 27, No. 3 (July 1948), pp. 379-423 (Part I), pp. 623-56 (Part II).
4. D. A. Huffman, "A Method for the Construction of Minimum Redundancy Codes," *Proc. IRE*, 40 (September 1952), pp. 1098-101.
5. R. Hunter and A. H. Robinson, "International Digital Facsimile Coding Standards," *Proc. IEEE*, 68, No. 7 (July 1980), pp. 830-46.
6. H. Gharavi and A. N. Netravali, "CCITT Compatible Coding of Multi-level Pictures," 62, No. 9, Part 1 (November 1983), pp. 2765-78.

AUTHOR

Hamid Gharavi, B.S.E.E., 1970, Tehran Polytechnic; M.S.C., 1975 (Digital Communication), and Ph.D. (Electrical Engineering), 1979, University of Technology, Loughborough, England; University of Technology, 1979-1980; Auckland University, New Zealand, 1980-1981; AT&T Bell Laboratories, 1982-1983. Present affiliation Bell Communications Research, Inc. Before joining AT&T Bell Laboratories, Mr. Gharavi was a Research Fellow at the

University of Technology and a Lecturer at Auckland University. Mr. Gharavi has worked on problems related to digital modulations for satellite communication. He has also worked on bandwidth compression of color television signals, source coding, graphics, and pattern recognition. Member, IEEE.